



ECE317 : Feedback and Control

Lecture : Relative stability

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Course roadmap



Modeling

- ✓ Laplace transform
- ✓ Transfer function
- ✓ Block Diagram
- ✓ Linearization
- ✓ Models for systems
 - electrical
 - mechanical
 - example system

Analysis

- ✓ Stability
 - Pole locations
 - Routh-Hurwitz
- ✓ Time response
 - Transient
 - Steady state (error)
- ✓ Frequency response
 - Bode plot

Design

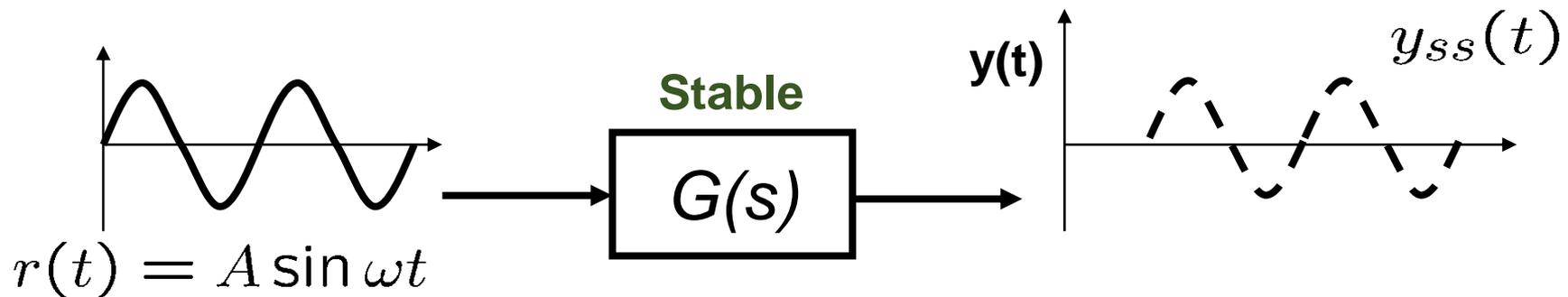
- Design specs
- Frequency domain
- Bode plot
- Compensation
- Design examples

Matlab & PECS simulations & laboratories

Frequency response (review)



- Steady state output $y_{ss}(t) = A |G(j\omega)| \sin(\omega t + \angle G(j\omega))$
 - **Frequency** is same as the input frequency ω
 - **Amplitude** is that of input (A) multiplied by $|G(j\omega)|$
 - **Phase** shifts $\angle G(j\omega)$
- Gain**

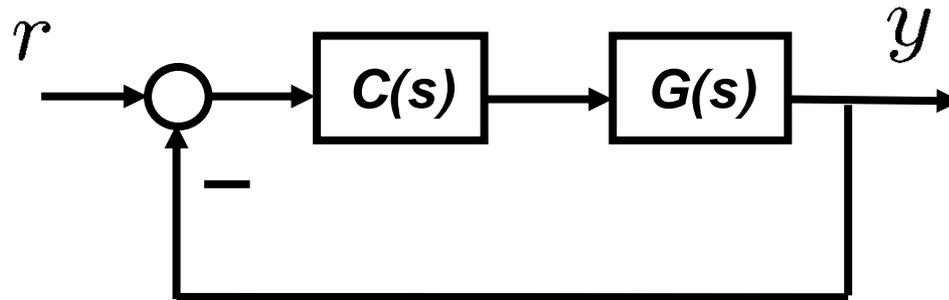


- **Frequency response function** (FRF): $G(j\omega)$
- **Bode plot**: Graphical representation of $G(j\omega)$

Stability of feedback system



- Consider the feedback system



- Fundamental questions
 - If G and C are stable, is the closed-loop system *always stable*?
 - If G and C are unstable, is the closed-loop system *always unstable*?

Closed-loop stability criterion



- Closed-loop stability can be determined by the roots of the **characteristic equation**

$$1 + L(s) = 0, \quad L(s) := G(s)C(s)$$

- Closed-loop system is stable if the Ch. Eq. has all roots in the open left half plane.
- How to check the closed-loop stability?
 - Computation of all the roots
 - Routh-Hurwitz stability criterion
 - **Relative stability criterion (phase margin):** *Open-loop FRF $L(j\omega)$ contains information of closed-loop stability.*

Advantages of using frequency response to determine stability



- It does not require transfer functions, just **experimental frequency response data** of the (stable) open-loop system are necessary to judge the closed-loop stability. On the other hand, Routh-Hurwitz criterion needs transfer functions.
- It leads to the concept of “**stability margin**”, i.e., gain-margin and phase-margin. From Routh-Hurwitz criterion, we can only judge “stable or not”.

Remarks on stability margin criterion



- Stability margin criterion gives not only *absolute* but also *relative stability*.
 - **Absolute stability**: Is the closed-loop system stable or not? (Answer is yes or no.)
 - **Relative stability**: How “much” is the closed-loop system stable? (Margin of safety)
- Relative stability (stability margin) is important because a math model is never accurate.
- How to measure relative stability?
 - Gain margin (GM) & Phase margin (PM)

Gain margin (GM)



- Phase crossover frequency ω_p :

$$\angle L(j\omega_p) = -180^\circ$$

- **Gain margin** (in dB)

$$GM = 20 \log_{10} \frac{1}{|L(j\omega_p)|}$$

- Indicates how much OL gain can be multiplied without violating CL stability.

Phase margin (PM)



- Gain crossover frequency ω_g :

$$|L(j\omega_g)| = 1$$

- Phase margin

$$PM = \angle L(j\omega_g) + 180^\circ$$

- Indicates how much OL phase lag can be added without violating CL stability.

Phase margin test for stability



- (Under a some conditions*) Closed loop stability of a system is guaranteed when

Phase margin is positive ($PM > 0$)

i.e. the phase of the system needs to be greater than -180 degrees at the gain crossover frequency

- * i) there is exactly one gain crossover frequency
- ii) the system is open-loop stable

Phase margin test for stability

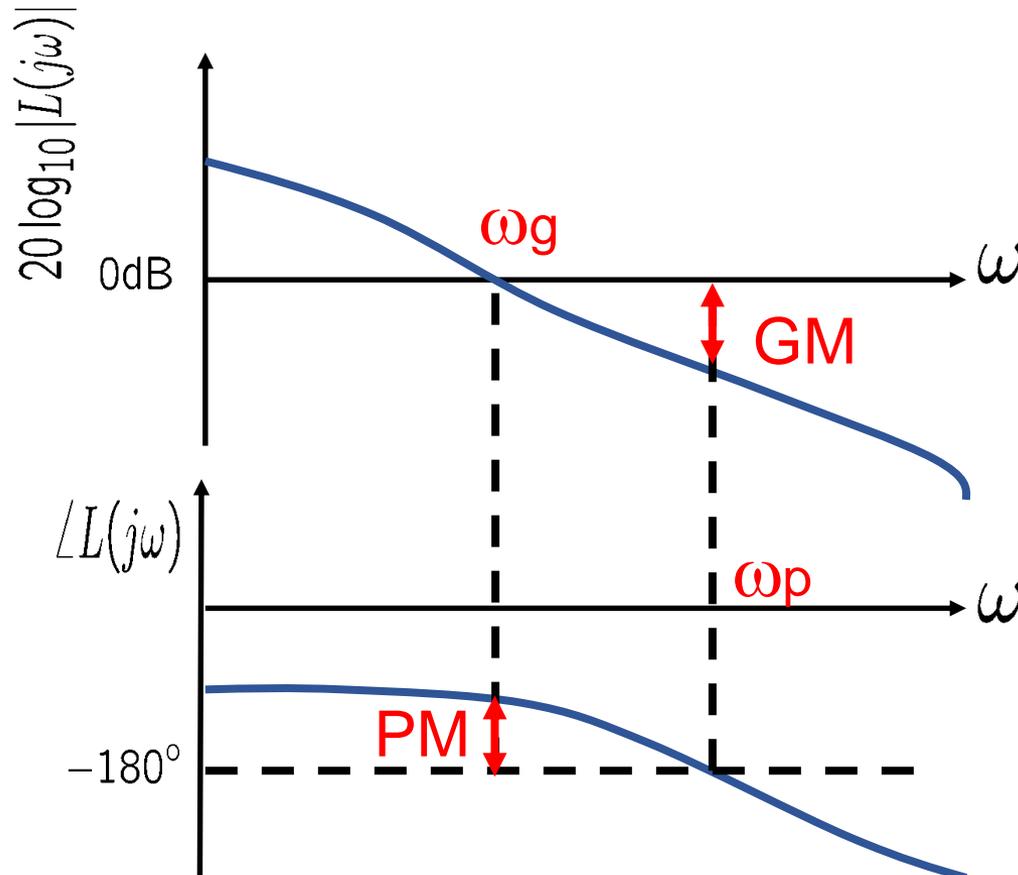


- Note under this test there is NO requirements for gain margin.
- However, it is generally stated that gain margin must also be positive. This can be shown to not be true by a counter example.

Relative stability on Bode plot



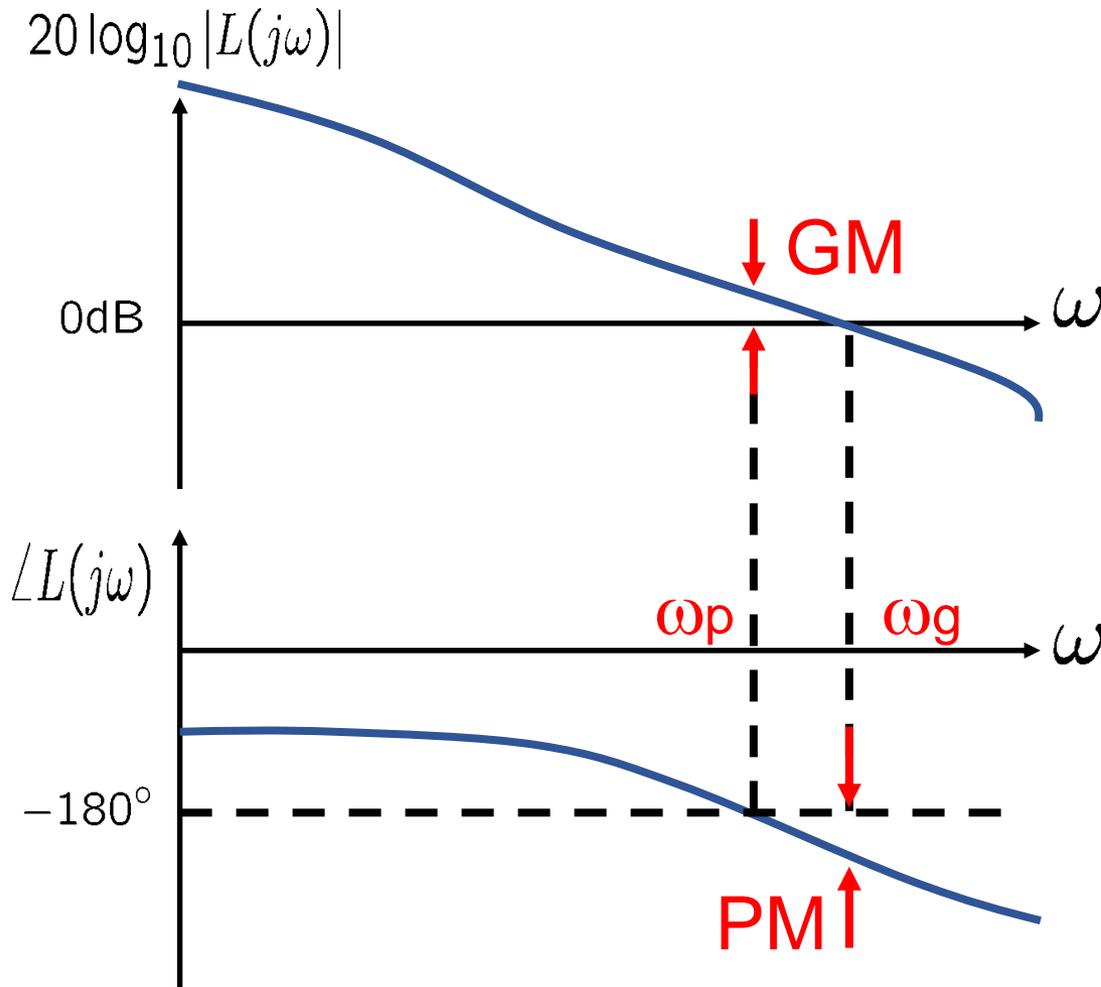
- When $\angle L(j\omega_g) > -180^\circ \rightarrow$ **PM** is positive otherwise, it is negative
- When $|L(j\omega_p)| < 0dB \rightarrow$ **GM** is positive, otherwise it is negative



Here **GM** is positive

Here **PM** is positive

Unstable closed-loop case



Here **GM** is negative

Here **PM** is negative

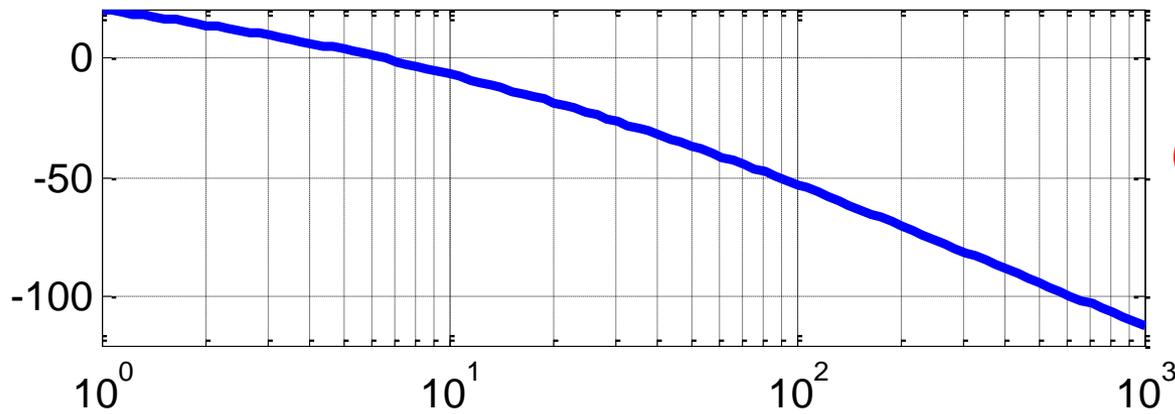
Notes on Bode plot



- Advantages
 - Without computer, Bode plot can be sketched easily by using straight-line approximations.
 - GM, PM, crossover frequencies are easily determined on Bode plot.
 - Controller design on Bode plot is simple.
- Disadvantage
 - If OL system has poles in open right half plane, it will be complicated to use Bode plot for closed-loop stability analysis.

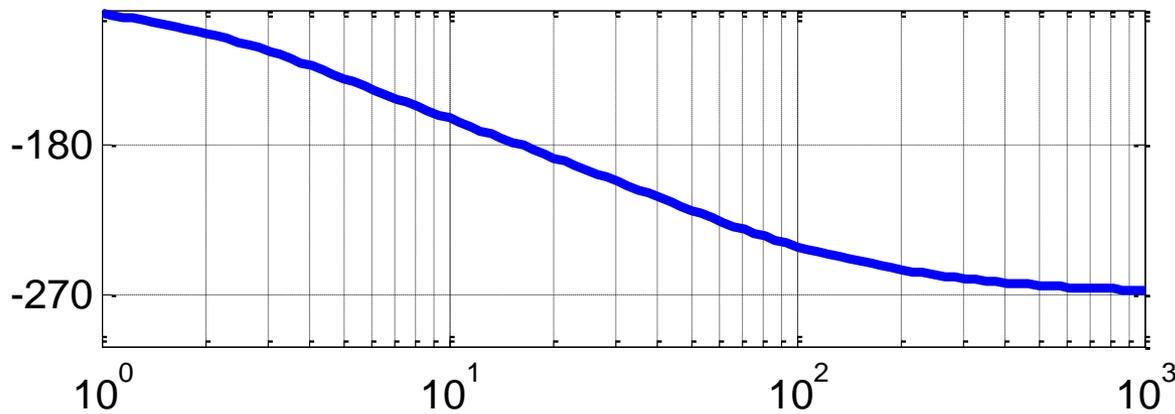
Example 1:

$$L(s) = \frac{2500}{s(s+5)(s+50)}$$



$\omega_g = ?$

GM: positive? negative?

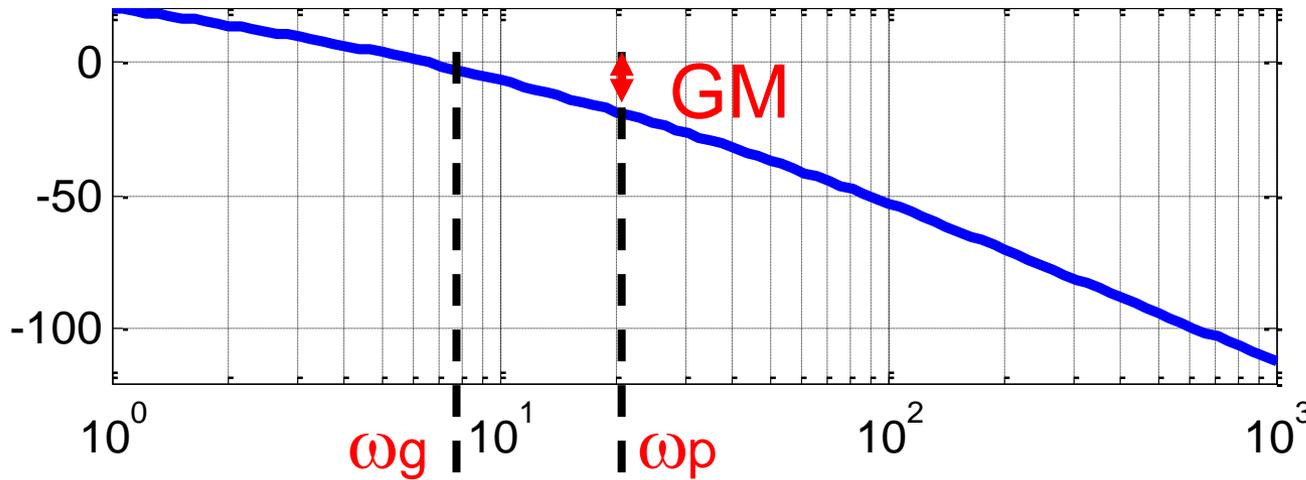


$\omega_p = ?$

PM: positive? negative?

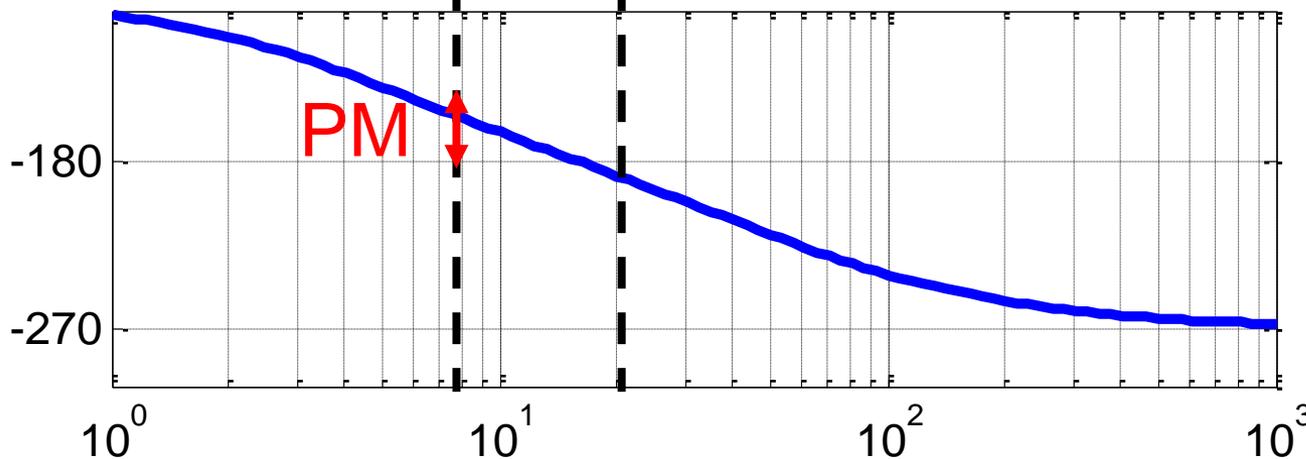
Example, cont'd:

$$L(s) = \frac{2500}{s(s+5)(s+50)}$$



$\omega_p = 16$ rds/s

GM: positive



$\omega_g = 6$ rds/s

PM: positive

\Rightarrow closed-loop system is STABLE

Example, cont'd:

$$L(s) = \frac{2500}{s(s+5)(s+50)}$$



- Using straight line asymptotic approximation determine:
 - i. Unity gain crossover frequency: ω_g
 - ii. Phase margin: **PM**
 - iii. -180 degree phase crossover frequency: ω_p
 - iv. Gain margin: **GM**

- confirm the results with Matlab 'margin' command

(This example is worked out in class and homework)

Example 2:



Sketch the asymptotic Bode plot for the following loop gain.

Annotate the plots completely:

- 1) Show the values of all break frequencies for magnitude and phase,
- 2) For magnitude plots: show i) gain along all straight line segments, and ii) slopes,
- 3) For phase plots: show the slopes.

$$L(s) = \frac{A}{\left(1 + \frac{s}{\omega_0}\right) \left(1 + \frac{s}{\omega_1}\right)}$$

where:

$$A = 200, \quad \omega_0 = 100, \quad \omega_1 = 300$$

(This example is worked out in class and homework)

Example 3:



Sketch the asymptotic Bode plot for the following loop gain.

Annotate the plots completely and **sketch using frequency in Hz** (not rds/s):

- 1) Show the values of all break frequencies for magnitude and phase,
- 2) For magnitude plots: show i) gain along all straight line segments, and ii) slopes,
- 3) For phase plots: show the slopes.

$$L(s) = \frac{A \left(1 - \frac{s}{\omega_z}\right)}{1 + \frac{s}{Q\omega_0} + \left(\frac{s}{\omega_0}\right)^2}$$

where:

$$A = 120, \quad \omega_z = 2\pi(2500), \quad \omega_0 = 2\pi(500), \quad Q = 5$$

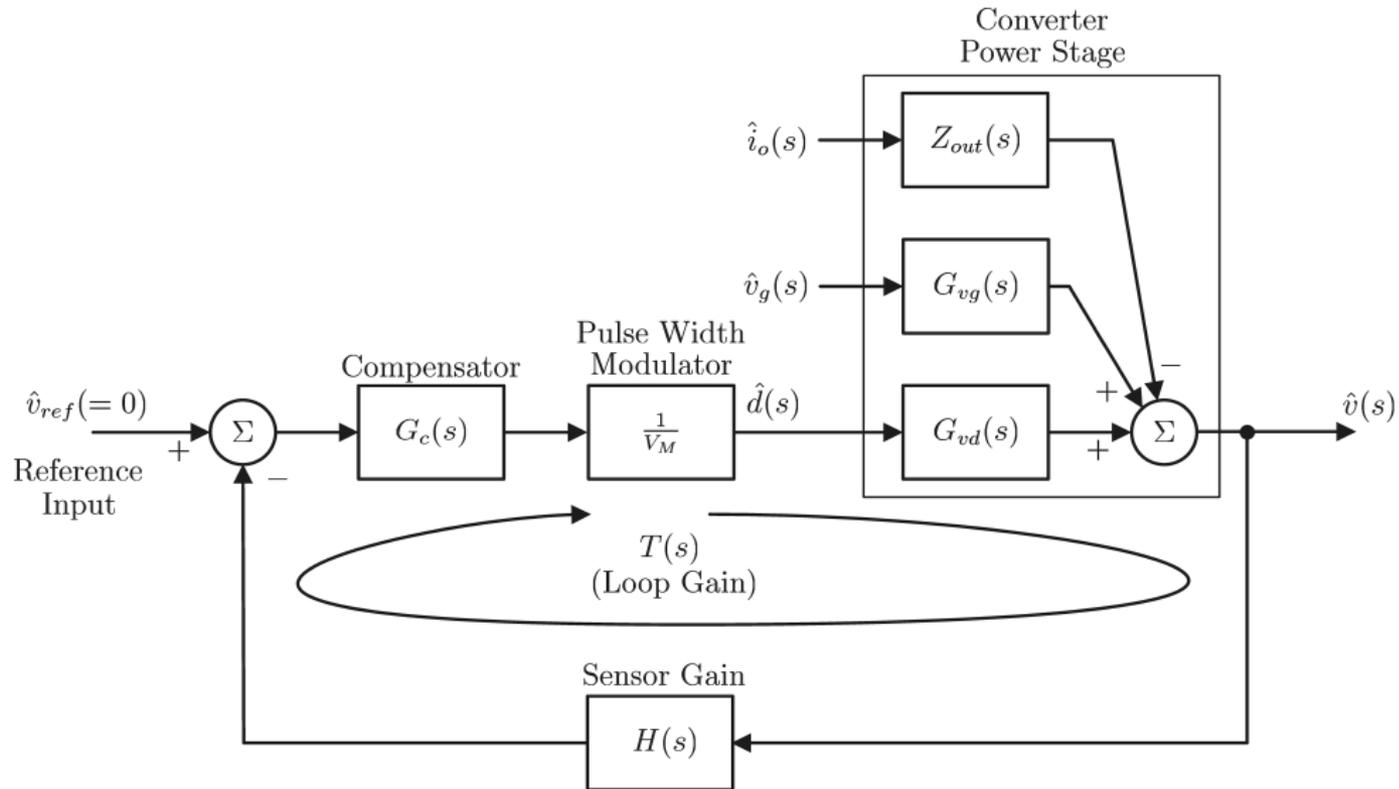
(This example is worked out in class and homework)

Summary



- Relative stability:
 - Gain margin, phase crossover frequency
 - Phase margin, gain crossover frequency
- Relative stability on Bode plot
- We normally emphasize PM in controller design.

Application to the lab:



$$T(s) = \frac{1}{V_M} \cdot G_c(s) \cdot G_{vd}(s) \cdot H(s)$$

In the lab $T(s)$ is used to refer to the loop gain $L(s)$

Application to the lab. Cont'd



Block diagram reduction leads to the closed loop transfer functions:

$$\hat{v} = G_{vref_CL}(s)\hat{v}_{ref} + G_{vg_CL}(s)\hat{v}_g - Z_{out_CL}(s)\hat{i}_o$$

$$G_{vref_CL}(s) = \frac{1}{H(s)} \frac{T(s)}{1 + T(s)}$$

$$G_{vg_CL}(s) = \frac{G_{vg}(s)}{1 + T(s)}$$

$$Z_{out_CL}(s) = \frac{Z_{out}(s)}{1 + T(s)}$$

where:

$$T(s) = \frac{1}{V_M} \cdot G_c(s) \cdot G_{vd}(s) \cdot H(s)$$

Application to the lab. Cont'd



$$T(s) = \frac{1}{V_M} \cdot G_c(s) \cdot G_{vd}(s) \cdot H(s)$$

- To **determine absolute stability** of this system we can use Routh-Hurwitz criterion. Note however, this is NOT applied to $T(s)$, but rather the Routh-Hurwitz criterion is applied to the denominator polynomial of

$$\frac{1}{1 + T(s)} \quad \text{or} \quad \frac{T(s)}{1 + T(s)}$$

- To determine **absolute stability and relative stability** of the system we find the phase and gain margins exhibited by the loop gain $T(s)$